

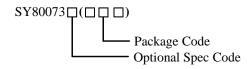
$21\mu A$ Ultra Low I_Q, 1.5MHz, 3A Synchronous Step Down Regulator

General Description

The SY80073 is a high efficiency 1.5MHz synchronous step-down DC/DC regulator capable of delivering up to 3A output current. It can operate over a wide input voltage range from 2.5V to 5.5V and integrates main switch and synchronous switch with very low $R_{DS\;(ON)}$ to minimize the conduction loss.

The SY80073 is in a space saving, low profile SOT563 package.

Ordering Information



Ordering Number	Package Type	Note
SY80073ART	SOT563	

Features

- 2.5V to 5.5V Input Voltage Range
- Ultra-Fast Load Transient Speed
- Low $R_{DS(ON)}$ for Internal Switches (Top/Bottom) $45m\Omega/35m\Omega$
- High Switching Frequency 1.5MHz Minimizes the External Components
- ±1.0% Feedback or Output Voltage Accuracy Over 0°C~ 85°C Temperature Range
- ±1.5% Feedback or Output Voltage Accuracy Over -40°C ~125°C Temperature Range
- PFM Mode for Light Load Efficiency
- 21µA Operating Quiescent Current
- Internal Soft-Start Limits the Inrush Current
- 100% Dropout Operation
- Output Auto Discharge Function
- Auto Recovery for SCP/OVP/OTP Protection
- RoHS Compliant and Halogen Free
- Compact Package: SOT563

Applications

- Portable Electronics
- Industrial PC
- Smart Phone

Typical Applications

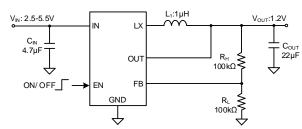


Figure 1. Schematic Diagram

Inductor and C_{OUT} Selection Table

V _{OUT} (V)	L(µH)	C _{OUT} (µF)			
		10	22	44	
1.2	0.68		☆	$\sqrt{}$	
1.2	1.0		$\sqrt{}$	\checkmark	
1.8	0.68		V		
	1.0		☆	$\sqrt{}$	
2.5	0.68		V		
	1.0		☆	$\sqrt{}$	

Note: '☆' means recommended for most applications.

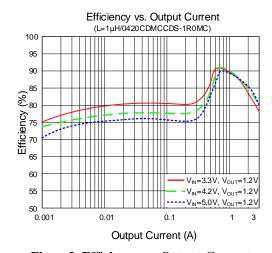
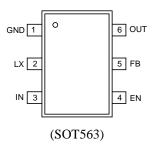


Figure 2. Efficiency vs. Output Current



Pinout (top view)



Top Mark: d3xyz (Device code: d3, x=year code, y=week code, z=lot number code)

Pin Name	Pin Number	Pin Description			
GND	1	Power ground pin.			
LX	2	Inductor pin. Connect this pin to the switching node of inductor.			
IN	3	nput pin. Decouple this pin to the GND pin with at least a 4.7μF ceramic capacitor.			
EN	4	Enable control. Pull high to turn on. Do not leave it floating.			
FB 5		Output feedback pin. Connect this pin to the center point of the output resistor			
ГБ	3	divider (as shown in Figure 1) to program the output voltage: $V_{OUT}=0.6\times(1+R_H/R_L)$.			
OUT	6	Output feedback pin, connected to the output capacitor side.			

Absolute Maximum Ratings (Note 1)

Supply Input VoltageFB, EN, OUT Voltage	$-0.3V$ to $V_{IN} + 0.6V$
Power Dissipation, PD @ TA = 25°C	
Package Thermal Resistance (Note 2)	
heta _{JA}	70°C/W
heta JC	
Junction Temperature Range	
Lead Temperature (Soldering, 10 sec.)	260°C
Storage Temperature Range (*1) LX Voltage Tested Down to -3V <20ns (*2) LX Voltage Tested Up to +7V <20ns	65°C to 150°C

Recommended Operating Conditions (Note 3)

Supply Input Voltage		2.5V to 5.5V
Ambient Temperature Range	;	-40°C to 85°C



Electrical Characteristics

 $(T_J=-40^{\circ}C\sim125^{\circ}C)$, and $V_{IN}=2.5V$ to 5.5V. Typical values are at $T_J=25^{\circ}C$ and $V_{IN}=5V$, unless otherwise specified. The values are guaranteed by test, design or statistical correlation.)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Voltage Range	V_{IN}		2.5		5.5	V
Input UVLO Threshold	$V_{\rm UVLO}$	V _{IN} falling	2.1	2.2	2.3	V
Input UVLO Hysteresis	V_{HYS}			160		mV
Quiescent Current	I_Q	$V_{FB}=105\%\times V_{REF}$		21		μA
Shutdown Current	I_{SHDN}	$V_{EN}=0V$, $T_{J}=25$ °C		0.05	0.5	μA
Feedback Reference Voltage	V_{REF}	I _{OUT} =1A, CCM	0.591	0.6	0.609	V
Output Discharge Resistance	R _{DIS}	$V_{EN} = 0V, V_{LX} = 0.4V$		8		Ω
Top FET R _{ON}	$R_{DS(ON)1}$			45		m Ω
Bottom FET Ron	R _{DS(ON)2}			35		mΩ
EN Input Voltage High	$V_{EN,H}$		1.1			V
EN Input Voltage Low	$V_{\text{EN,L}}$				0.4	V
EN Leakage Current	I _{EN,LKG}	EN=high		0.01	1	μA
Min ON Time	t _{ON,MIN}			100		ns
Maximum Duty Cycle (Note 4)	D_{MAX}		100			%
Soft Start Time	tss	from EN high to 95% of V _{OUT} nominal, T _J =25°C	400	600	800	μs
Switching Frequency	f_{SW}	$I_{OUT}=1A$, $V_{OUT}=1.8V$		1.5		MHz
Top FET Current Limit	I _{LMT,TOP}		4		6	A
Bottom FET Current Limit	I _{LMT,BOT}		3			A
Thermal Shutdown Temperature (Note 4)	T _{SD}			150		°C
Thermal Shutdown Hysteresis (Note 4)	T _{HYS}			20		°C

Note 1: Stresses beyond the "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note 2: Package thermal resistance is measured in the natural convection at T_A=25°C on a 6cm×6cm size 2-oz two-layer Silergy Evaluation Board.

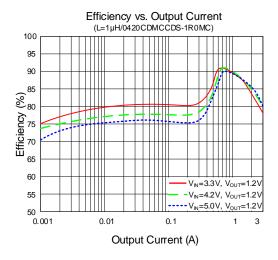
Note 3: The device is not guaranteed to function outside its operating condition.

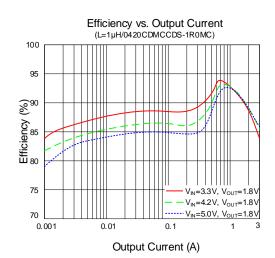
Note 4: Guaranteed by design.

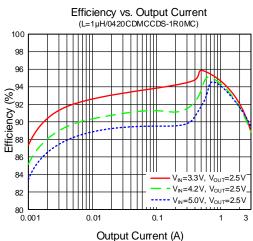


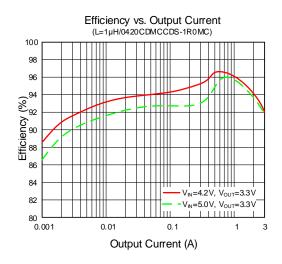
Typical Performance Characteristics

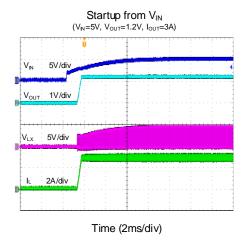
 $(T_A=25$ °C, $V_{IN}=5V$, $L=1\mu H$, $C_{OUT}=2\times 10\mu F$, unless otherwise specified.)

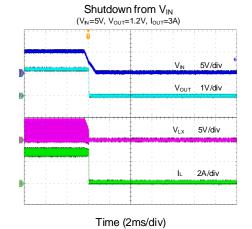




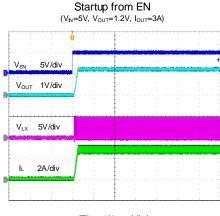










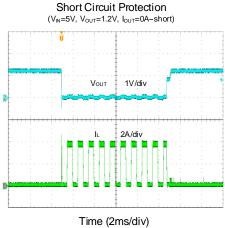


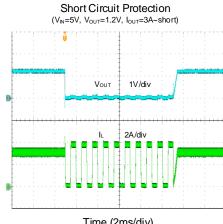


(V_{IN} =5V, V_{OUT} =1.2V, I_{OUT} =3A) 5V/div V_{OUT} 1V/div V_{LX} 5V/div 2A/div

Shutdown from EN

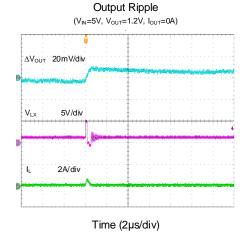
Time (2ms/div)

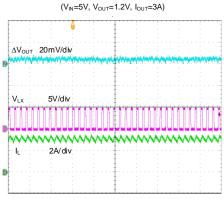




Time (2ms/div)

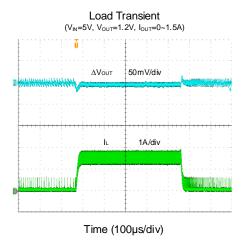
Output Ripple

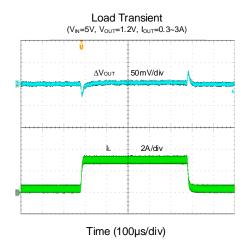


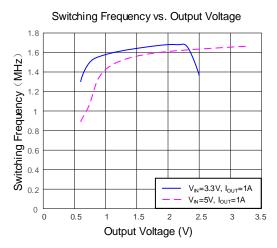


Time (2µs/div)











Operation

The SY80073 is a high efficiency 1.5MHz synchronous step-down DC/DC regulator capable of delivering up to 3A output current. It can operate over a wide input voltage range from 2.5V to 5.5V and integrates main switch and synchronous switch with very low $R_{\rm DS(ON)}$ to minimize the conduction loss.

The SY80073 adopts the instant PWM architecture to achieve fast transient responses for high step down applications and high efficiency at light loads.

The SY80073 provides protection functions such as cycle by cycle current limiting and thermal shutdown protection.

Low output voltage ripple, small external inductor and capacitor sizes are achieved with 1.5MHz switching frequency.

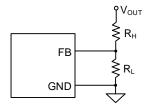
Applications Information

Because of the high integration in the SY80073, the application circuit based on this regulator is rather simple. Only the input capacitor $C_{\rm IN}$, the output capacitor $C_{\rm OUT}$, the output inductor L and the feedback resistors ($R_{\rm H}$ and $R_{\rm L}$) need to be selected for the targeted application specifications.

Feedback Resistor Dividers R_H and R_L:

Choose R_H and R_L to program the proper output voltage. To minimize the power consumption under light loads, it is desirable to choose large resistance values for both R_H and R_L . A value between $10k\Omega$ and $1M\Omega$ is highly recommended for both resistors. If V_{OUT} is 1.8V, $R_H=100k\Omega$ is chosen, then using following equation, R_L can be calculated to be $49.9k\Omega$:

$$R_{L} = \frac{0.6V}{V_{OUT} - 0.6V} R_{H}$$



Input Capacitor CIN:

The ripple current through input capacitor is calculated as:

$$I_{\text{CIN_RMS}} = I_{\text{OUT}} \times \sqrt{D(1 \text{-}D)}$$

To minimize the potential noise problem, place a typical X5R or better grade ceramic capacitor with 6.3V rating really close to the IN and GND pins. Care should be taken to minimize the loop area formed by C_{IN} , and IN/GND pins. In this case, a $4.7\mu F$ low ESR ceramic capacitor is recommended.

Output Capacitor Cour:

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For the best performance, it is recommended to use an X5R or better grade ceramic capacitor with 6.3V rating and more than $2\times10\mu F$ capacitance.

Output Inductor L:

There are several considerations in choosing this inductor.

1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum output current. The inductance is calculated as:

$$L = \frac{V_{\text{OUT}}(1 - V_{\text{OUT}}/V_{\text{IN,MAX}})}{f_{\text{SW}} \times I_{\text{OUT,MAX}} \times 40\%}$$

where f_{SW} is the switching frequency and $I_{OUT,MAX}$ is the maximum load current.

The SY80073 is quite tolerant of different ripple current amplitude. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

 The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

Isat, min > Iout, max +
$$\frac{\text{Vout}(1-\text{Vout/Vin,max})}{2 \times \text{fsw} \times L}$$

3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with DCR<25m Ω to achieve a good overall efficiency.



Inductor vs. Output Capacitor:

The ripple base control strategy needs very little C_{OUT} to confirm stability. Too large inductor and C_{OUT} will lead to unstability. The recommend inductance and output capacitance is shown as below.

Inductance vs. Output Capacitance Selection Table

L	C _{OUT} (μF)					
(µH)	22	44	88	120	180	220
0.68	Note5	√	√	$\sqrt{}$	$\sqrt{}$	$\sqrt{}$
1.0	V	V	√	√	×	×

Note 5: Only suitable for V_{OUT}<2.0V application.

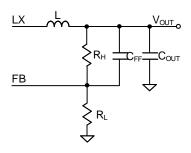
Minimum Duty Cycle and Maximum Duty Cycle:

In the COT architecture, there is no limitation for small duty cycles, since even at very low duty cycles, the switching frequency can be reduced as needed once the on-time is close to the minimum on time, to always ensure a proper operation.

The device will enter 100% dropout mode if the output voltage is very close to the input voltage, while the top FET is constantly turned on, the bottom FET is turned off.

Load Transient Considerations:

The SY80073 integrates the compensation components to achieve good stability and fast transient responses. Adding a small ceramic cap in parallel with $R_{\rm H}$ may further speed up the load transient responses and is thus recommended for applications with large load transient step requirements.



OCP and UVP Protection Method:

The SY80073 adopts cycle by cycle current limitation for both HS FET and LS FET. If the high side power FET current gets higher than the peak current limit threshold, the high side power FET will turn off and the low side power FET will turn on. If the low side FET current gets higher than the valley current limit threshold, the low side FET will keep turning on until low side FET current decreases below the valley current limit threshold. So, both peak and valley current are limited. If the load current continues to increase in these conditions, the output voltage will drop. When the output voltage falls below 33% of the regulation level, UVP will be triggered and the IC will operate in hic-cup mode. The hic-cup on time and hic-cup off time ratio is 1:1. If the hard short is removed, the IC will return to normal operation.

Layout Design:

The layout design of the SY80073 is relatively simple. For the best efficiency and minimum noise problem, the following components should be placed close to the IC: C_{IN}, L, R_H, R_L and C_{FF}.

- 1) It is desirable to maximize the PCB copper area connecting to GND pin to achieve the best thermal and noise performance. If the board space allowed, a ground plane is highly desirable. Place some vias in GND copper for heat sinking too.
- 2) C_{IN} must be close to Pins IN and GND. The loop area formed by C_{IN} and GND must be minimized.
- 3) The PCB copper area associated with the LX pin must be minimized to avoid the potential noise problem.
- 4) The components R_{H} and R_{L} , and the trace connecting to the FB pin must not be adjacent to the LX net on the PCB layout to avoid the noise problem.
- 5) The feedback sampling point should be connected to C_{OUT} rather than the inductor output terminal.
- 6) If the system chip interfacing with the EN pin has a high impedance state at shutdown mode and the IN pin is connected directly to a power source such as a Li-Ion battery, it is desirable to add a pull-down $1M\Omega$ resistor between the EN and GND pins to prevent the noise from falsely turning on the regulator at shutdown mode.



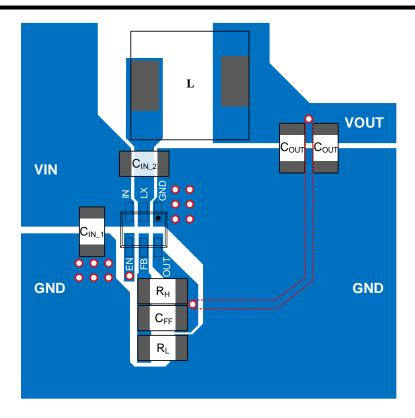
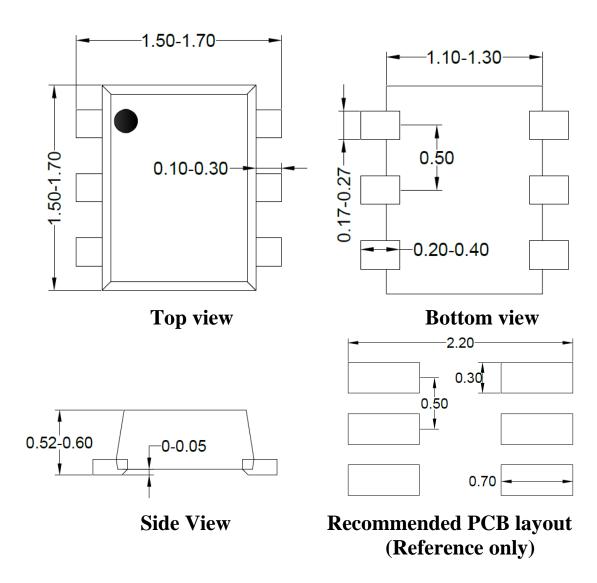


Figure 3. PCB Layout Suggestion



SOT563 Package Outline Drawing



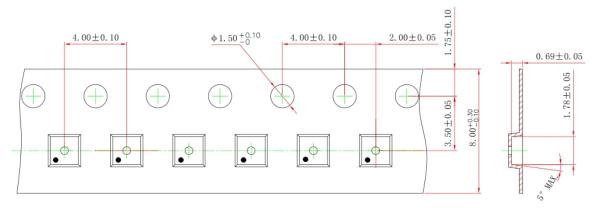
Notes: All dimension in millimeter and exclude mold flash & metal burr.



Taping & Reel Specification

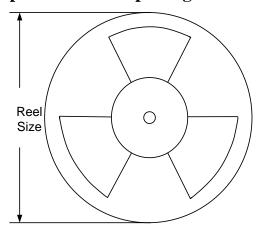
Taping Orientation

SOT563



Feeding Direction-

2. Carrier Tape & Reel specification for packages



Package types	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Trailer * length(mm)	Leader * length (mm)	Qty per reel (pcs)
SOT563	8	4	7''	280	160	5000

3. Others: NA



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